

1 Spectral Analysis of Bivariate Stationary Process

Definition of a Stationary Bivariate Process:

$$fX_{1,t}, fX_{2,t}, t: 0, 1, 2, \dots$$

$fX_{1,t}, X_{2,t}$ is bivariate stationary if and only if:

- a) $fX_{1,t}, fX_{2,t}$ are univariate stationary
- b) $\text{cov}(X_{1,t}, X_{2,t})$ is a function of $(s_j - t)$ only.

Let both processes be null mean.

$$\begin{aligned} R_{11}(s) &= E \left[X_{1,t} X_{1,t+s} \right] \\ R_{22}(s) &= E \left[X_{2,t} X_{2,t+s} \right] \\ R_{12}(s) &= E \left[X_{1,t} X_{2,t+s} \right] \\ R_{21}(s) &= E \left[X_{2,t} X_{1,t+s} \right] \end{aligned}$$

so,

$$R_{12}(s) = R_{21}(s)$$

The corresponding auto- correlations are:

$$\gamma_{11}(s) = \frac{R_{11}(s)}{R_{11}(0)}$$

$$\gamma_{22}(s) = \frac{R_{22}(s)}{R_{22}(0)}$$

$$\gamma_{12}(s) = \frac{R_{12}(s)}{[R_{11}(0)R_{22}(0)]^{1/2}}$$

similarly for $\frac{1}{2}_{21}(z)$:

$$\mathbf{R}(s) = \begin{pmatrix} R_{11}(s) & R_{12}(s) \\ R_{21}(s) & R_{22}(s) \end{pmatrix}$$

$$\frac{1}{2}\mathbf{Q}(s) = \begin{pmatrix} \frac{1}{2}_{11}(s) & \frac{1}{2}_{12}(s) \\ \frac{1}{2}_{21}(s) & \frac{1}{2}_{22}(s) \end{pmatrix}$$

The above are the covariance and correlation matrices; there are a countable infinity of them.

$$\begin{aligned} & \frac{1}{2}_{ii}(s) \leq 1, \quad \max \text{ at } s = 0 \\ & \frac{1}{2}_{ij}(s) = \frac{1}{2}_{ji}(s), \quad \text{symmetric in } s \\ & \frac{1}{2}_{ij}(s) \leq 1 \end{aligned}$$

but $\max \frac{1}{2}_{ij}(s)$ can be anywhere; $\frac{1}{2}_{ij}(s)$ is **not** symmetric in s .

Diagram

1.1 Auto and Cross Spectra

Provided the auto and cross covariances are absolutely summable;

i.e. provided

$$\sum_{s=i-1}^{\infty} |R_{ij}(s)| < 1$$

$$i, j = 1; 2$$

then the non-normalized spectra are:

$$h_{ij}(\omega) = \frac{1}{2\pi} \sum_{s=i-1}^{\infty} R_{ij}(s) e^{i\omega s}$$

For $i \neq j$ this is the **cross**-spectral density function; for $i = j$, we have the **auto**-spectral density function.

Reinterpreting this result:

By univariate stationarity:

$$X_{1,t} = \int_{-\infty}^t e^{i\omega(t-s)} dZ_1(s)$$

$$X_{2,t} = \int_{-\infty}^t e^{i\omega(t-s)} dZ_2(s)$$

$dZ_i(s)$ are each orthogonal processes.

Substitute the definitions of $fX_{1,t}$, $fX_{2,t}$ into the definition for $R_{12}(s)$:

$$R_{12}(s) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} e^{i\omega_1 t} e^{i(\omega_2 + t)\omega} E \int_{-\infty}^{\infty} f(\omega_1(\omega)) dZ_2^{\omega}(\omega) g$$

$R_{12}(s)$ is a function of s only by assumed stationarity; therefore, so must $R_{11}(s)$ be.

$R_{11}(s) = R_{22}(s)$ as a function of s if, and only if: $E \int_{-\infty}^{\infty} f(\omega_1(\omega)) dZ_2^{\omega}(\omega) g = 0$

$\delta_{ij} \neq 0$, i.e. Σ_{z_1, z_2} is a diagonal matrix

Bivariate stationary processes can be represented by orthogonal random process that that are also mutually orthogonal.

$$h_{ii}(\omega) = E \int_{-\infty}^{\infty} j dZ_i(\omega) j^2$$

as before. But also:

$$h_{12}(\omega) = E \int_{-\infty}^{\infty} f(\omega_1(\omega)) dZ_2^{\omega}(\omega) g$$

$h_{ii}(\omega)$ contains the square of the coefficients of $e^{i\omega t}$ in $X_{i,t}$

$h_{ij}(\omega)$ contains the product of the coefficients of $e^{i\omega t}$ in $X_{1,t}$ and $X_{2,t}$

$h_{ii}(\omega)$ gives the contribution to **total variance** of frequency ω .

$h_{12}(\omega)$ gives the contribution to "total covariance" of frequency ω .

The spectral density matrix can be related by inverse F:T: to the covariance matrix.

$$h_{ij}(\omega) = \frac{1}{2\pi} \int_{s=-\infty}^{\infty} R_{ij}(s) e^{i\omega s} ds$$

$i, j = 1, 2$

Similarly:

$$f_{ij}(\omega) = \frac{1}{2\pi} \int_{s=-\infty}^{\infty} \frac{1}{2} R_{ij}(s) e^{i\omega s} ds$$

1.2 Various Complex Decompositions of $h_{12}(\omega)$

$h_{12}(\omega)$ is complex in general, even for real processes.
We rewrite:

$$h_{12}(\omega) = \underbrace{\text{Re}\{h_{12}(\omega)\}}_{\text{co-spectrum}} + i \underbrace{\text{Im}\{h_{12}(\omega)\}}_{\text{quadrature spectrum}}$$

real component imaginary component

By replacing $e^{i\omega s}$ by $\cos(\omega s) - i \sin(\omega s)$ in

$$h_{12}(\omega) = \frac{1}{2\pi} \int_{s=-\infty}^{\infty} R_{12}(s) e^{i\omega s} ds$$

we see that

$$G_{12}(\omega) = \frac{1}{2} \int_{-\infty}^{\infty} [R_{12}(s) + R_{12}^*(s)] \cos(\omega s) ds$$

$$Q_{12}(\omega) = \frac{1}{2} \int_{-\infty}^{\infty} [R_{12}(s) - R_{12}^*(s)] \sin(\omega s) ds$$

$\frac{1}{2} [R_{12}(s) + R_{12}^*(s)]$ is the **even** part of $R_{12}(s)$
 $\frac{1}{2} [R_{12}(s) - R_{12}^*(s)]$ is the **odd** part of $R_{12}(s)$;
 i.e. we have separated the non-symmetric $R_{12}(s)$
 into the sum of an **even** part and an **odd** part.

At $\omega = 0$:

$$h_{12}(0) = G_{12}(0)$$

$$Q_{12}(0) = 0$$

1.3 Polar Form Expressions

$$h_{12}(\omega) = \mathcal{R}_{12}(\omega) e^{i\tilde{A}_{12}(\omega)}$$

$$\mathcal{R}_{12}(\omega) = |h_{12}(\omega)| = \sqrt{G_{12}^2 + Q_{12}^2}$$

$$\tilde{A}_{12}(\omega) = \tan^{-1} \frac{Q_{12}(\omega)}{G_{12}(\omega)}$$

$\mathcal{R}_{12}(\omega) \gg$ cross-amplitude spectrum

$\tilde{A}_{12}(\omega) \gg$ phase spectrum

$$h_{12}(\omega) = \int_{-\infty}^{\infty} f(z_1(\omega)) dz_2^*(\omega) g$$

Let

$$dZ_i(\omega) = j dZ_i(\omega) j e^{i\tilde{A}_i(\omega)}$$

$$i = 1, 2$$

Therefore

$$h_{12}(\omega) = \frac{E \int dZ_1(\omega) j dZ_2(\omega) j g}{E e^{i[\tilde{A}_1(\omega) + \tilde{A}_2(\omega)]}}$$

Therefore

$$\tilde{h}_{12}(\omega) = \frac{E \int dZ_1(\omega) j dZ_2(\omega) j g}{E e^{i[\tilde{A}_1(\omega) + \tilde{A}_2(\omega)]}}$$

Remember $\tilde{A}_{12}(\omega)$ is only defined mod 2π .

Complex coherency is a "standardization" of the complex cross spectrum.

$$!_{12}(\omega) = \frac{h_{12}(\omega)}{[h_{11}(\omega) h_{22}(\omega)]^{1/2}}$$

and can be interpreted as the **correlation coefficient** between $fX_{1,t}, fX_{2,t}$ at **frequency !**

$$0 \leq !_{12}(\omega) \leq 1$$

$!_{12}(\omega) \gg$ a correlation coefficient in frequency domain

But remember, in general, $!_{12}(\omega)$ is **complex**.

$!_{12}^2(\omega)$ is **invariant** to linear transformations for **stationary** processes only- but there can be a change in phase relationships.

Let $fX_{1,t}, fX_{2,t}$ be any two time series that are

bivariate stationary.

Let

$$\begin{aligned}
 X_{1,t} &= \sum_{u=1}^t a(u) X_{1,t-u} + \epsilon_{1,t} \\
 X_{2,t} &= \sum_{u=1}^t b(u) X_{2,t-u} + \epsilon_{2,t}
 \end{aligned}$$

Therefore

$$\begin{aligned}
 dZ_1(t) &= \epsilon_{1,t} \\
 dZ_2(t) &= \epsilon_{2,t}
 \end{aligned}$$

Therefore

$$\begin{aligned}
 R_{11}(t) &= \epsilon_{1,t}^2 \\
 R_{22}(t) &= \epsilon_{2,t}^2 \\
 R_{12}(t) &= \epsilon_{1,t} \epsilon_{2,t}
 \end{aligned}$$

And so,

$$(R_{12}(t))^2 = (R_{11}(t) R_{22}(t))$$

where

$$R_{12}(t) = \frac{R_{12}(t)}{\sqrt{R_{11}(t) R_{22}(t)}}$$

Therefore,

$$j\omega_{12}(\omega)j^2 = \frac{\int_{-\infty}^{\infty} e^{i\omega t} \Phi_{22}^{-1}(\omega) h_{12}^2(\omega) d\omega}{\int_{-\infty}^{\infty} e^{i\omega t} j^2 h_{11}(\omega) h_{22}(\omega) d\omega}$$

$$= j\omega_{12}(\omega)j^2$$

$j\omega_{12}(\omega)j = 1$ if and only if $fX_{1,t}$ is a linear transform of $fX_{2,t}$.

Let

$$X_{1,t} = \sum_{j=1}^n \tilde{A}_j X_{2,t} e^{i\omega_j t}, \quad \sum_{j=1}^n \tilde{A}_j^2 < 1$$

Therefore

$$X_{1,t} = \sum_{j=1}^n \tilde{A}_j e^{i\omega_j t} \int_{-\infty}^{\infty} e^{i\omega t} dZ_2(\omega)$$

so,

$$dZ_1(\omega) = \sum_j \tilde{A}_j e^{i\omega_j t} dZ_2(\omega)$$

but this is a functional linear relationship in frequency space, so,

$$j\omega_{12}(\omega)j^2 = 1$$

Some Examples:

1) **Uncorrelated** (in time) process:

i.e.

$$\begin{aligned}
 & R_{12}(s) = 0, & \text{ss} \\
 \mathbf{V} & h_{12}(\omega) = 0, & \text{gl} \\
 \mathbf{V} & \tilde{R}_{12}(\omega) = 0 \\
 & C_{12}(\omega) = \alpha_{12}(\omega) = 0, & \text{gl} \\
 & \beta_{12}(\omega) = 0
 \end{aligned}$$

2) Linearly (structurally) related:

$$X_{1,t} = aX_{2,t} + \epsilon_t$$

compare example above.

ϵ_t is white noise, uncorrelated with $X_{2,t}$
therefore

$$R_{12}(s) = aR_{22}(s)$$

and

$$\begin{aligned}
 h_{12}(\omega) &= ah_{22}(\omega) \\
 h_{11}(\omega) &= a^2h_{22}(\omega) + h_{\epsilon\epsilon}(\omega)
 \end{aligned}$$

so,

$$C_{12}(\omega) = ah_{22}(\omega); \quad \alpha_{12}(\omega) = 0$$

and

$$\tilde{R}_{12}(\omega) = ah_{22}(\omega); \quad \tilde{A}_{12}(\omega) = 0$$

therefore

$$\beta_{12}(\omega) = \frac{h_{12}(\omega)}{[h_{11}(\omega)h_{22}(\omega)]^{\frac{1}{2}}}$$

$$\begin{aligned}
&= \frac{P_{h_{22}}(\omega)}{P_{a^2 h_{22}(\omega) + h_{\epsilon}}(\omega)} \\
&= \frac{1}{1 + \frac{h_{\epsilon}(\omega)^{3/4}}{a^2 h_{22}(\omega)^{1/2}}}
\end{aligned}$$

If ϵ is white noise, $h_{\epsilon}(\omega)$ is a **constant**.

1.4 Regression with Delay- the most interesting case

Let

$$X_{1,t} = aX_{2,t-1} + \epsilon_t$$

therefore

$$R_{12}(s) = aR_{22}(s-i)$$

and as

$$h_{12}(\omega) = \frac{1}{2^{1/4}} \sum_{s=i-1}^{\infty} R_{12}(s) e^{i\omega s},$$

we get:

$$\begin{aligned}
h_{12}(\omega) &= \frac{a}{2^{1/4}} \sum_{s=i-1}^{\infty} R_{22}(s-i) e^{i\omega s} \\
&= a e^{i\omega} h_{22}(\omega)
\end{aligned}$$

$$h_{11}(\omega) = a^2 h_{22}(\omega) + h_{\epsilon}(\omega)$$

$$G_{12}(\omega) = a \cos \left(\frac{\omega z}{v} \right) H_{22}(\omega)$$

Note the increase
in frequency

$$G_{12}(\omega) = a \sin \left(\frac{\omega z}{v} \right) H_{22}(\omega)$$

$$\tilde{A}_{12}(\omega) = a H_{22}(\omega)$$

$$\tilde{A}_{12}(\omega) = \frac{d \tilde{A}_{12}(\omega)}{d \omega}$$

Particularly important!

A time delay \Rightarrow phase spectrum is a linear function of frequency and the **slope** is the magnitude of the **delay**.

A generalization of this is to consider for any phase spectrum:

$$\tau(\omega) = - \frac{d \tilde{A}_{12}(\omega)}{d \omega}$$

is known as the "envelope, group delay."

1.5 A More Complex Example with Feedback Loop (Like a cob web cycle)

$$X_2(t) = a X_1(t - \tau) + \dots \quad X_1(t) = b X_2(t) + \dots$$

Note delay

w_t, \hat{w}_t are independent white noise processes with spectral densities:

$$h_{w_t}(\omega) = \frac{\frac{3}{4}^2}{2 \cdot \frac{1}{4}} \quad h_{\hat{w}_t}(\omega) = \frac{\frac{3}{4}^2}{2 \cdot \frac{1}{4}}$$

Let us use the orthogonal components approach.

$$dZ_2(\omega) = a e^{i\omega t} dZ_1(\omega) + dZ_w(\omega)$$

$$dZ_1(\omega) = j b dZ_2(\omega) + dZ_{\hat{w}}(\omega)$$

Solve for dZ_1, dZ_2 :

$$dZ_1(\omega) = \frac{dZ_{\hat{w}}(\omega) + j b dZ_w(\omega)}{(1 + a b e^{i\omega t})}$$

$$dZ_2(\omega) = \frac{a e^{i\omega t} dZ_{\hat{w}}(\omega) + dZ_w(\omega)}{(1 + a b e^{i\omega t})}$$

Use the result that:

$$h_{ii}(\omega) = \int dZ_i(\omega) j^2$$

$i = 1, 2$

$$h_{11}(\omega) = \frac{\frac{3}{4}^2}{2 \cdot \frac{1}{4}} + b^2 \frac{\frac{3}{4}^2}{2 \cdot \frac{1}{4}}}{j(1 + a b e^{i\omega t}) j^2}$$

$$h_{22}(\omega) = \frac{\frac{1}{2 \cdot \frac{1}{4}} a^2 \frac{3}{4}^2 + \frac{3}{4}^2}{j(1 + a b e^{i\omega t}) j^2}$$

$$h_{12}(\theta) = \frac{\frac{1}{2} [ae^{i\theta/4} + b^{3/4}]}{j(1 + abe^{i\theta})}$$

$$|1 + abe^{i\theta}|^2 = 1 + a^2b^2 + 2ab\cos(\theta)$$

$$h_{11}(\theta) = \frac{a^{3/4} + b^{3/4}}{2(1 + a^2b^2 + 2ab\cos(\theta))}$$

$$h_{22}(\theta) = \frac{a^{3/4} + b^{3/4}}{2(1 + a^2b^2 + 2ab\cos(\theta))}$$

$$h_{12}(\theta) = \frac{ae^{i\theta/4} + b^{3/4}}{2(1 + a^2b^2 + 2ab\cos(\theta))}$$

$$G_2(\theta) = \frac{a^{3/4}\cos(\theta) + b^{3/4}}{2(1 + a^2b^2 + 2ab\cos(\theta))}$$

$$Q_2(\theta) = \frac{a^{3/4}\sin(\theta)}{2(1 + a^2b^2 + 2ab\cos(\theta))}$$